





Solving the constrained output regulation problem using projected dynamical systems

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Constrained Output Regulation. Given a plant P_0 , a class of references \mathcal{R} , and a set V, solve (when possible) the following output regulation problem:



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Motivations. Guarantee safety constraints and avoid the windup phenomenon. Our focus will be on integral control, thus $\mathcal{R} = \text{constant references}$.

Projected Integral Controllers (Lorenzetti & Weiss, 2022)



Keywords Safety Constraints, Projected Dynamical Systems, Singular Perturbations.

P. Lorenzetti and G. Weiss. "PI control of stable nonlinear plants using projected dynamical systems," Automatica, 2022. 2

A. Integral Control

A brief review on integral control for linear and nonlinear systems.



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B. Projected Dynamical Systems

How is the operator Π_U defined?



Integral Control

Integral Control for Linear Systems (Davison, 1976; Morari, 1985)



Theorem. Let **P** be a linear plant. Assume that:

- (a) \mathbf{P} is stable and $\mathbf{P}(0)$ is full row rank.
- (b) $\sigma(\mathbf{P}(0)K) \subset \mathbb{C}_+$, i.e., $-\mathbf{P}(0)K$ Hurwitz.

Then, there exists a $\kappa > 0$ such that for all $k \in (0, \kappa)$ the above closed-loop system

is stable and it exhibits zero tracking error for all constant references r.

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Possible choice. Choose K such that $K = \mathbf{P}(0)_{\text{right}}^{-1}$.

Integral Control for Nonlinear Systems (Desoer & Lin, 1985)



Theorem. Let \mathbf{P} be a nonlinear system. Assume that:

- (a) For every constant input $u \in \mathbb{R}^m$, $\Xi(u)$ is a **uniform GES** equilibrium point of **P**.
- (b) The map G is strictly monotone for all $v_1, v_2 \in \mathbb{R}^m$.

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Linear Systems. (a) $\sigma(A) \subset \mathbb{C}_{-}$; (b) $\mathbf{P}(0) + \mathbf{P}(0)^{\top}$ strictly positive definite.

The Idea Behind: Singular Perturbations



boundary-layer (fast) system



(a)

reduced (slow) dynamics



Projected Dynamical Systems

Normal and Tangent Cones

Definition. Let $U \subset \mathbb{R}^m$ be a closed, non-empty, convex set. We define

$$\begin{split} N_{U}(u) &= \{ z \in \mathbb{R}^{m} \mid \langle z, v - u \rangle \leq 0 \quad \forall \ v \in U \} & \forall \ u \in U, \\ N_{U}(u) &= \emptyset & \forall \ u \notin U. \end{split}$$
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Definition. Let $U \subset \mathbb{R}^m$ be a closed, non-empty, convex set. For every $u \in U$ and $w \in \mathbb{R}^m$, we define

$$\Pi_U(u,w) = \arg\min_{v \in T_U(u)} \|w - v\|.$$



Definition. Let $F \in C(\mathbb{R}^m; \mathbb{R}^m)$ and $U \subset \mathbb{R}^m$ be non-empty, closed and convex. A function $u \in W^{1,1}((0,\tau); \mathbb{R}^m)$ that satisfies

$$\dot{u}(t) = \prod_{U} (u(t), -F(u(t)))$$
 $u(0) = u_0$ (PDS)

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How to safely regulate the output power in a grid-connected synchronverter? 4. Future Perspectives

What are possible future directions?

Control Problem Formulation

The nonlinear plant \mathbf{P}_0 to be controlled is described by:

$$\dot{x} = f_0(x, v), \qquad y = g(x),$$

with $f_0 \in C^2(\mathbb{R}^n \times \mathbb{R}^n; \mathbb{R}^n)$, $g \in C^1(\mathbb{R}^n; \mathbb{R}^p)$, with $m \ge p$.

Control Objective

The control objective is to make the output signal y tracks a **constant reference** signal $r \in Y \subset \mathbb{R}^p$, while making sure that the plant input signal v remains in a **desired compact set** $V \subset \mathbb{R}^m$ (e.g., determined by operational constraints).

Closed-Loop System - Equations



The closed-loop system is described by

$$\dot{x} = f_0(x, \mathcal{N}(u_I)), \quad \dot{u}_I = \prod_U (u_I, k(r - g(x))), \tag{CL}$$

where $U \subset \mathbb{R}^p$ is compact and convex, $\mathcal{N} \in C^2(\mathbb{R}^p, \mathbb{R}^m)$, $V = \mathcal{N}(U)$, and k > 0.

Closed-Loop Stability Analysis

Assumption 1. There exists an open domain $\mathcal{V} \subset \mathbb{R}^m$ and $\Xi \in C^1(\mathcal{V}; \mathbb{R}^n)$ such that

$$f_0(\Xi(v), v) = 0 \qquad \forall v \in \mathcal{V},$$

and the equilibrium points $\{\Xi(v) | v \in \mathcal{V}\}$ are uniformly locally exponentially stable.

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Linear Systems. Let \mathbf{P}_0 be linear $(\dot{x} = Ax + Bv, y = Cx)$, Assumption 1 reduces to A being **stable**. The function Ξ is given by

$$\Xi(v) = (-A)^{-1}Bv.$$

Assumption 2. P₀ satisfies Assumption 1. Let $G(v) := g(\Xi(v)) \in C^1(\mathcal{V}; \mathbb{R}^p)$. There exist an open set $\mathcal{U} \subset \mathbb{R}^p$, a function $\mathcal{N} \in C^2(\mathcal{U}, \mathcal{V})$, and $\mu > 0$ such that

$$\langle G(\mathcal{N}(u_1)) - G(\mathcal{N}(u_2)), u_1 - u_2 \rangle \ge \mu ||u_1 - u_2||^2$$

for all $u_1, u_2 \in \mathcal{U}$, i.e., $G \circ \mathcal{N}$ is strictly monotone.

Choice of \mathcal{N} . When possible, we suggest the choice $\mathcal{N} = G_{\text{right}}^{-1}$, so that $\mathcal{U} = G(\mathcal{V})$.

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Linear Systems. Let \mathbf{P}_0 be linear and $\mathcal{N} \in \mathbb{R}^{m \times p}$, Assumption 2 reduces to $\mathbf{P}_0(0)\mathcal{N} + (\mathbf{P}_0(0)\mathcal{N})^{\top}$ being **strictly positive definite**. The function *G* is given by $G(v) = C\Xi(v) = C(-A)^{-1}Bv = \mathbf{P}_0(0)v.$

Mappings Recap



Notation. We denote $Y = G(\mathcal{N}(U))$, and, for any $r \in Y$, we define

$$u_r := (G \circ \mathcal{N})^{-1}(r)$$
 $x_r := \Xi(\mathcal{N}(u_r)).$

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Theorem 1. Consider the closed-loop system (CL), where \mathbf{P}_0 satisfies Assumption 2. Then there exists a $\kappa > 0$ such that if the gain $k \in (0, \kappa]$, then for any $r \in Y$, $(\Xi(\mathcal{N}(u_r)), u_r)$ is a (locally) **exponentially stable equilibrium point** of the closed-loop system (CL), with state space $X = \mathbb{R}^n \times \mathcal{U}$. If the initial state $\begin{bmatrix} x_0 \\ u_0 \end{bmatrix} \in X$, of the closed-loop system satisfies $u_0 \in U$ and $\|x_0 - \Xi(\mathcal{N}(u_0))\| \le \varepsilon_0$, then

 $x(t) \to \Xi(\mathcal{N}(u_r)), \qquad u_I(t) \to u_r, \qquad y(t) \to r,$

and this convergence is at an exponential rate.

Intuition of the Result



Reduced (slow) Model

Boundary-Layer (fast) System

$$\xrightarrow{u_I} \mathcal{N} \xrightarrow{v} \mathbf{P_0} \xrightarrow{y}$$

 $\dot{x} = f_0(x, \mathcal{N}(u_I)),$

where $u_I \in U$ is **fixed**.

$$u_{I} \xrightarrow{G \circ \mathcal{N}} y$$

$$\int \Pi_{U} \xrightarrow{-r}$$

$$\frac{\mathrm{d}u_I}{\mathrm{d}s} = \Pi_U(u_I, r - G(\mathcal{N}(u_I))),$$

where $s = k \cdot t$ slow time-scale.

Constrained Power Regulation for a Grid-Connected Synchronverter

A Grid-Connected Inverter with an LC Filter



The above inverter is controlled as a **synchronverter**, i.e., as a **synchronous generator**.

P. Lorenzetti, Z. Kustanovich, S. Shivratri and G. Weiss, "The equilibrium points and stability of grid-connected synchronverters," *IEEE Transactions on Power Systems*, 2022.

Closed-Loop Control Scheme



Assumption 1 - The set \mathcal{V} ($\Xi: \mathcal{V} ightarrow \mathbb{R}^4$)



Assumption 2 - The sets ${\cal U}$ and U when ${\cal N}=\overline{G}_{
m right}^{-1}$



Assumption 2 - The sets \mathcal{U} when $\mathcal{N} = K \in \mathbb{R}^{2 \times 2}$



Classical I control loop



Projected I control loop



Numerical Results - The Integrator State u_I and the Set U



Numerical Results - The Plant Input v and the Set V



Conclusion & Perspectives

Projected Integrator

Projected Systems Π_U Safety Constraints Stability Analysis Singular Perturbations Gain $\mathcal{N}=G_{\mathrm{right}}^{-1}$

Power Systems Power Regulation for a Synchronverter

Time-Varying Constraints. In some power systems applications, we would like

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Weaker Notions of Stability. What can we say on aGAS systems?

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Infinite-Dimensional Systems. Preliminary results for well-posed SISO linear systems.

Thanks for your attention!